

# Performance evaluation of an adaptive backoff scheme for WLAN

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## Summary

In this paper, a simple self-adaptive contention window adjustment algorithm for 802.11 wireless local area networks (WLAN) is proposed and analyzed. Numerical results show that the new algorithm outperforms the standard 802.11 window adjustment algorithm. Compared with the standard and previously proposed enhancement algorithms, a salient feature of our algorithm is that it performs well in both heavy and light contention cases regardless of the packet sizes and physical versions. Moreover, the adaptive window adjustment algorithm is simpler than previously proposed schemes in that no live measurement of the WLAN traffic activity is needed. Copyright © 2004 John Wiley & Sons, Ltd.

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**KEY WORDS:** IEEE 802.11; WiFi; wireless local area networks; WLAN; CSMA/CA; self-adaptive contention window adjustment

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## 1. Introduction

The wide-spread adoption of 802.11 wireless local area networks (WLANs) operating at the unlicensed ISM frequency bands [1] is one of the few highlights of communication technologies in recent years. The IEEE 802.11 standard [2] specifies the medium access control (MAC) and physical (PHY) layers for so called 802.11 WLANs. Within the family of 802.11 WLANs, the most widely deployed version so far is 802.11b, which operates at 2.4 GHz and provides up to 11 Mbps data rate. Another standardized version is 802.11a, which operates at 5 GHz and provides up to 54 Mbps data rate. The newest version, 802.11g,

which operates at 2.4 GHz as well but provides 54 Mbps data rate, has been finalized in June 2003.

Two types of MAC protocols, distributed coordination function (DCF) and point coordination function (PCF), are defined in 802.11. Most commercial products only implement DCF. The DCF mechanism is simple and robust. However, it has been shown by many that the standard DCF cannot efficiently utilize the limited wireless channel bandwidth when there are many stations in the WLAN accessing the same channel [3–9]. The major reason is that the initial contention window size is kept fixed regardless of the traffic activity, whereas ideally it should be large when the number of active stations is large and *vice versa*.

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The major contribution of this paper is a novel self-adaptive contention-window adjustment algorithm—MIMLD (multiplicative increase multiplicative/linear decrease) algorithm. Unlike the standard algorithm, this algorithm automatically adjusts the initial contention window to a near optimal point according to the traffic activity, thus avoiding bandwidth wastage due to improper contention window setting. Compared with other performance enhancement algorithms [3–8], our algorithm is effective not only when there are many active stations that contend with each other for channel access, but also when there are few stations. Furthermore, our algorithm does not need on-line measurement and computation.

The rest of the paper is organized as follows. Section 2 reviews the 802.11 standard and related work. Section 3 describes the proposed algorithm. Section 4 presents simulation and analytical results in various scenarios, and compares the performance of the new and standard algorithms. Section 5 concludes the paper.

## 2. 802.11 Standard and Related Work

DCF is the fundamental MAC layer function in 802.11 WLANs. DCF is based on carrier sense multiple access with collision avoidance (CSMA/CA). To resolve collisions of packets simultaneously transmitted by different stations, a slotted binary exponential backoff algorithm is employed in DCF.

At each packet transmission, a random backoff time (in slots) is selected uniformly between 0 and  $cw - 1$ . The value of  $cw$  is called the contention window. At the first transmission attempt,  $cw$  is set to a value  $CW_{min}$  called the minimum contention window. After each unsuccessful transmission,  $cw$  is doubled, up to a maximum value  $CW_{max}$ . After a successful transmission,  $cw$  will be reset to  $CW_{min}$  for the next packet. The value of  $CW_{min}$  is 16 and 32 respectively for 802.11a and 802.11b. The value of  $CW_{max}$  is 1024 in both cases<sup>‡</sup> [2].

The contention window adjustment algorithm defined in 802.11 has been proven to be robust in

simulations as well as in real applications. However, with a fixed  $CW_{min}$ , the original algorithm neglects the possibility that the number of actively contending stations can change dynamically over time, leading to dynamically changing contention intensity. When there are many active stations, too small a  $CW_{min}$  may lead to too many collisions and backoffs; on the other hand, when there are few active stations, too high a  $CW_{min}$  may lead to excessive idle airtime during which no station attempts to transmit. In either case, the channel is not used efficiently.

To solve the problem, one could employ a dynamic contention window adjustment algorithm with contention window adjusted to reflect the number of active stations in the WLAN. These contention window adjustment algorithms are different from the 802.11 WLAN standard and have been discussed in References [3–8]. Basically, these methods can be divided into two classes, measurement based and non-measurement based.

In the measurement-based methods [3–6], a relationship between the number of stations and the optimal contention window is first established. Then an on-line sniffer is built inside each station to monitor the activities of all surrounding stations. When the number of active stations is estimated, the contention window is adjusted accordingly. However, on-line measurement of active stations and computation of the optimal contention window incur extra processing cost and are hard to implement. Measurement and computation errors could even lead to worse performance than the original algorithm in the 802.11 standard. To make matters worse, the optimal contention window does not only depend on the number of active stations but also on the packet sizes. In real networks, where the applications generate traffic with a wide variety of packet sizes, these measurement-based methods have their limitations.

Two non-measurement-based methods were studied in References [7,8]. These methods do not adjust the values of control parameters (e.g.  $CW_{min}$ ) any more once they have been set up. They are easy to implement. But they only considered the case that there are a large number of stations. However, in real applications, it is quite common that there are only a few active stations with packets to send even if there are many stations in the WLAN. Web browsing, for example will only cause a client station to transmit packets sporadically. Also in the home environment, the number of stations itself is normally quite small.

Our goal is to find a non-measurement based scheme that is simple to implement and able to

<sup>‡</sup>In the 802.11 standard [2], 15, 31 and 1023 are actually used for  $CW_{min}$  and  $CW_{max}$ . The divergence comes from the different understanding of the value range of a new random backoff time, either  $[0, cw)$  or  $[0, cw]$ . For convenience here, we use 16, 32 and 1024 instead.

provide near-optimal performance in both light and heavy contention cases. Moreover, the scheme should perform well for all versions of PHY and for the full range of packet sizes. To achieve the target, in this paper a self-adaptive contention window algorithm—MIMLD that emulates the TCP window adjustment mechanism is proposed.

Our algorithm is in essence the generalization of algorithms in References [2,7,8]. The algorithms in References [2,7,8] are all special cases of our MIMLD algorithm. Compared with those in [2,7,8], our algorithm is more flexible and provides better performance over a wider range of scenarios. As will be demonstrated, the algorithm performs well both when the number of active stations is large and when it is small. And, as with TCP, it does not require direct measurement of the traffic activity in the channel.

### 3. Description of MIMLD Algorithm

Figure 1 and Figure 2 demonstrate the new self-adaptive contention window adjustment algorithm for WLAN. The new algorithm is simple in principle. The major difference between the new algorithm and the original 802.11 standard is that the initial contention window in the new algorithm adapts to the contention intensity of the wireless channel. The initial contention window of a packet is the contention window used for its first transmission.

At any one time, the window adjustment algorithm, MIMLD, can be in one of the three possible phases: multiplicative increase phase, multiplicative decrease phase and linear decrease phase. In the new algorithm, a new control parameter called  $CW_{basic}$  is introduced and the meaning of  $CW_{min}$  is changed slightly.  $CW_{basic}$  plays the role of a threshold for distinguishing the contention intensity of the wireless channel.  $CW_{basic}$  is typically set to be close to the value of  $CW_{min}$  in the original 802.11 algorithm.

- $cw \leftarrow \max(cw/2, CW_{basic})$ , if (succeeds and  $cw > CW_{basic}$ ).
- $cw \leftarrow \max(-cw, CW_{min})$ , if (succeeds and  $cw \leq CW_{basic}$ ).
- $cw \leftarrow \min\{2 \times \max(cw, CW_{basic}), CW_{max}\}$ , if collides.
- $cw \leftarrow cw$ , if retry limit is reached.

$$0 \leq CW_{min} \leq cw \leq CW_{max}$$

$$0 \leq CW_{min} \leq CW_{basic} \leq CW_{max}$$

Fig. 1. Multiplicative increase multiplicative/linear decrease (MIMLD) backoff algorithm.

When the contention window  $cw > CW_{basic}$ , we assume that the contention intensity in the wireless channel is high. If a packet is successfully transmitted when  $cw > CW_{basic}$ , instead of going back to  $CW_{min}$  immediately, the contention window is halved but lower-bounded by  $CW_{basic}$ . By setting the contention window at a relatively high level (relative to the original 802.11 algorithm), potential collisions in the future can be minimized. This phase is called ‘multiplicative decrease’.

When the contention window  $cw \leq CW_{basic}$ , after a successful transmission, the contention window is reduced by one rather than halved. Smaller contention window can yield better performance when there are a small number of active stations or the traffic is asymmetric (e.g. dominated by traffic from AP to clients). The intent of the linear reduction is to keep the contention window in the small regime as long as possible—reducing it too quickly may cause collisions to occur sooner, which in turn will cause  $cw$  to move out of this region. The linear reduction procedure is stopped when the contention window reaches its minimum value,  $CW_{min}$ . This phase is called ‘linear decrease’.

Recall that  $CW_{basic}$  is typically set to be close to the  $CW_{min}$  value of the original 802.11 algorithm. The reason why we allow the contention window to go below  $CW_{basic}$  is based on the observation that when contention is light (e.g. when there are only one or two active stations) it is not necessary to wait an average  $(CW_{basic} - 1)/2$  time slots before each transmission attempt. However, although the contention is assumed to be light when  $cw$  is below  $CW_{basic}$ , this region is also regarded as critical, since the contention window is small. To be conservative and to avoid oscillations, instead of continually decreasing the window multiplicatively, it is linearly decreased. Only several consecutive successful transmissions can lower the contention window to the minimum value  $CW_{min}$ .

When a collision occurs, if  $cw > CW_{basic}$ , the contention window is doubled as in the original algorithm. If  $cw < CW_{basic}$ , the contention window is immediately increased to  $CW_{basic} \times 2$ . The reason for that is to escape the critical area below  $CW_{basic}$ , where  $cw$  is small, to avoid the potential for more collisions. The idea is that  $cw$  below  $CW_{basic}$  should be used only when the number of active stations is small. The occurrence of a collision reduces the probability of that case. This window increase phase is called ‘multiplicative increase’.

By comparing MIMLD with the TCP congestion window adjustment procedure, similarities can be

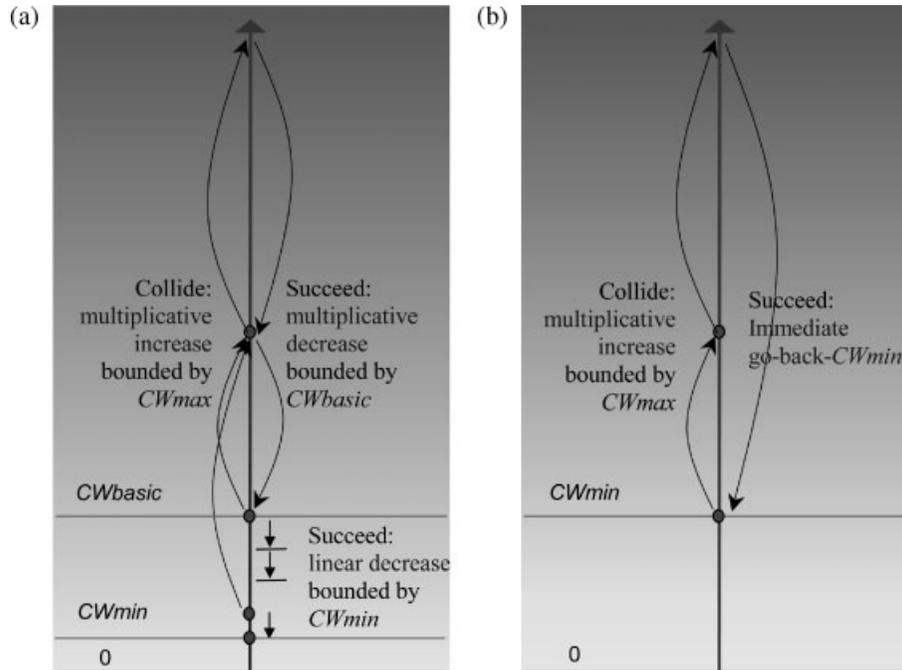


Fig. 2. Comparison of the contention window evolution processes in the two algorithms. (a) MIMLD algorithm. (b) Standard algorithm.

found. TCP throughput is proportional to its congestion window while 802.11 MAC throughput is proportional to the reciprocal of its contention window. The basic algorithms in TCP are additive increase and multiplicative decrease (AIMD) and 'slow' start [10]. Our 'multiplicative decrease' phase resembles the 'slow start', the 'linear decrease' phase resembles the 'additive increase' in TCP and the 'multiplicative increase' phase resembles the 'multiplicative decrease' algorithm in TCP (Table I).

MIMLD is simple to implement in that it does not require on-line sniffing and measurement of traffic activity. In addition, the intrinsic operation of MIMLD

takes into account both light and heavy contention cases. The next section shows that MIMLD algorithm improves the performance of the original 802.11 algorithm in both cases.

#### 4. Performance Evaluation

In this section, the performance of MIMLD is analyzed and compared with the standard algorithm. The analytical analysis is verified by simulation using NS-2 [11]. Table II shows the physical properties of 802.11a/11g and 802.11b in our study.<sup>§</sup>

Table I. Similarity between MIMLD and TCP.

MIMLD contention window evolution	TCP congestion window evolution
Multiplicative decrease	'Slow start'
Linear decrease	'Additive increase'
Multiplicative increase	'Multiplicative decrease'

<sup>§</sup>While in 802.11g standard SIFS is  $10\ \mu\text{s}$ , in practice it is effectively  $16\ \mu\text{s}$  because of a  $6\ \mu\text{s}$  signal extension that gets tacked on to the end of every OFDM frame. Effectively, in NS-2 simulation, not losing accuracy, a pure 802.11g environment is assigned the same parameters as that of 802.11a [12].

#### 4.1. Throughput and Delay Analysis

We assume that there are  $n$  contending stations in a WLAN. The buffers of all stations are saturated and the stations always have packets to send. The analytical evaluation of the MIMLD algorithm is given as follows. For the analysis of the standard algorithm, please refer to Reference [9].

Let  $p$  denote the probability of a packet colliding with another packet. Same as the model in Reference [9],

$$W_i = \begin{cases} 2^i W & 0 \leq i \leq m \\ W + i & -d \leq i < 0 \end{cases} \quad (1)$$

Consider the two-dimensional discrete-time Markov chain given by  $\{s(t), b(t)\}$ . By using the short notation,  $P\{i_1, k_1 | i_0, k_0\} = P\{s(t+1)=i_1, b(t+1)=k_1 | s(t)=i_0, b(t)=k_0\}$ , the only non-null one-step transition probabilities in the Markov chain are as follows (Figure 3):

$$\left\{ \begin{array}{l} P\{i, k | i, k+1\} = 1 \\ P\{i, k | i+1, 0\} = (1-p)/W_i \\ P\{i, k | i-1, 0\} = p/W_i \\ P\{1, k | i, 0\} = p/W_1 \\ P\{m, k | m, 0\} = p/W_m \\ P\{-d, k | -d, 0\} = (1-p)/W_{-d} \end{array} \right. \begin{array}{l} -d \leq i \leq m, 0 \leq k \leq W_i - 2 \\ -d \leq i \leq m-1, 0 \leq k \leq W_i - 1 \\ 1 \leq i \leq m, 0 \leq k \leq W_i - 1 \\ -d \leq i \leq -1, 0 \leq k \leq W_i - 1 \\ 0 \leq k \leq W_m - 1 \\ 0 \leq k \leq W_{-d} - 1 \end{array} \quad (2)$$

$p$  is assumed to be constant and independent of the backoff stage of the packet (i.e. independent of how many collisions the packet has already suffered). Let  $s(t)$  be the random process representing the backoff stage at time  $t$ . Let  $b(t)$  be the random process representing the backoff counter during a particular backoff stage. Other definitions of parameters are as follows:

- $\tau$ : the probability of a station transmitting during a slot time;
- $m$ : the maximum backoff stage beyond which the window size will be kept constant;
- $W$ : the basic window size, i.e. the value of  $CW_{\text{basic}}$ ;
- $d$ :  $-d$  is the minimum backoff stage; we have the relations  $d = W - CW_{\text{min}}$  and  $0 \leq d < W$ ;
- $W_i$ : the window size at backoff stage  $i$ ,  $-d \leq i \leq m$ ;

Table II. Physical properties of 802.11b, 802.11a and 802.11g.

	802.11b	802.11a/802.11g
Slot time	20 $\mu$ s	9 $\mu$ s
CCA time	15 $\mu$ s	3 $\mu$ s
RxTx turnaround time	5 $\mu$ s	2 $\mu$ s
SIFS time	10 $\mu$ s	16 $\mu$ s
PHY overhead	192 $\mu$ s	20 $\mu$ s
Data rate	11 Mbps	54 Mbps
Basic data rate	2 Mbps	6 Mbps

The first line in Equation (2) corresponds to the decrement of the backoff counter during the backoff process. The second line corresponds to the successful transmission of the previous packet and the new packet starts transmission. The third and fourth lines correspond to packet collision and the increment of the backoff stage. The fifth line corresponds to the fact that once the maximum stage is reached, there is no increment in subsequent transmissions. The sixth line corresponds to the fact that once the minimum stage is reached, there is no decrement in subsequent transmissions.

Let  $b_{i,k} = \lim_{t \rightarrow \infty} \Pr\{s(t) = i, b(t) = k\}$  be the stationary distribution of the Markov chain. We have,

$$\left\{ \begin{array}{l} b_{i,0} = \left(\frac{p}{1-p}\right)^{i-1} b_{1,0} \quad 1 \leq i \leq m \\ b_{i,0} = (1-p)^{-i+1} b_{1,0} \quad -d < i \leq 0 \\ b_{-d,0} = \frac{(1-p)^{d+1}}{p} b_{1,0} \end{array} \right. \quad (3)$$

and,

$$b_{i,k} = \frac{W_i - k}{W_i} b_{i,0}, \quad -d \leq i \leq m, \quad 1 \leq k \leq W_i - 1 \quad (4)$$

By imposing the normalization condition for stationary probability distribution, we have,

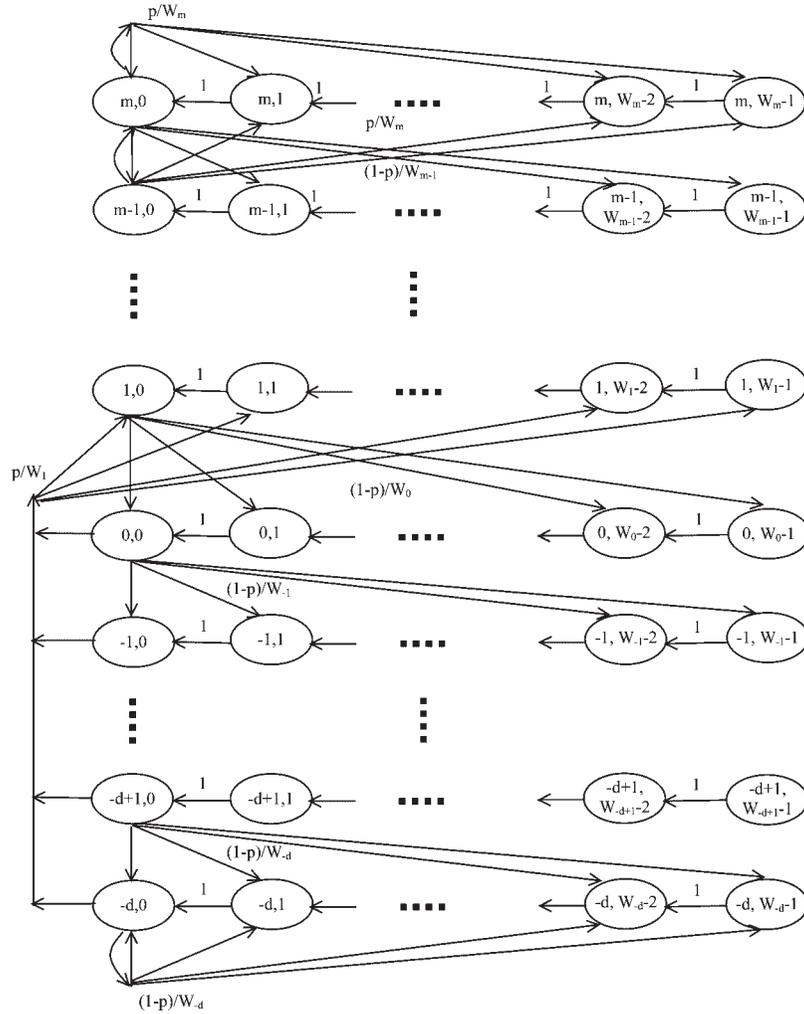


Fig. 3. State transition diagram of the new algorithm.

$$\begin{aligned}
 1 &= \sum_{i=-d}^m \left( b_{i,0} \sum_{k=0}^{W_i-1} \frac{W_i - k}{W_i} \right) = \sum_{i=-d}^m \left( b_{i,0} \frac{W_i + 1}{2} \right) \\
 &= \frac{b_{1,0}}{2} \left[ \sum_{i=1}^m \left( \frac{p}{1-p} \right)^{i-1} (W_i + 1) + \sum_{i=-(d-1)}^0 (1-p)^{-i+1} (W_i + 1) + \frac{(1-p)^{d+1}}{p} (W_{-d} + 1) \right] \\
 &= \frac{b_{1,0}}{2} \left[ \left( \sum_{i=1}^m 2W \left( \frac{2p}{1-p} \right)^{i-1} + \sum_{i=1}^m \left( \frac{p}{1-p} \right)^{i-1} \right) \right. \\
 &\quad \left. + \left( \sum_{i=-(d-1)}^0 (W + i + 1)(1-p)^{-i+1} + (W - d + 1) \frac{(1-p)^{d+1}}{p} \right) \right] \\
 &= \frac{b_{1,0}}{2} [\Delta_1 + \Delta_2]
 \end{aligned} \tag{5}$$

$\Delta_1$  and  $\Delta_2$  in Equation (5) are derived as follows.

$$b_{1,0} = \frac{2}{[\Delta_1 + \Delta_2]}$$

$$= \frac{2}{[(2W(1 - (2q)^m)(1 - q) + (1 - 2q)(1 - q^m)) / ((1 - 2q)(1 - q) + (1 - p)(Wp + 2p - 1 + (1 - p)^{d+1})) / p^2]}$$

For simplicity, let  $q = p/(1 - p)$ ,

$$\begin{aligned} \Delta_1 &= \left( \sum_{i=1}^m 2W(2q)^{i-1} + \sum_{i=1}^m q^{i-1} \right) \\ &= 2W \left( \frac{1 - (2q)^m}{1 - 2q} \right) + \frac{1 - q^m}{1 - q} \\ &= \frac{2W(1 - (2q)^m)(1 - q) + (1 - 2q)(1 - q^m)}{(1 - 2q)(1 - q)} \\ \Delta_2 &= \sum_{i=-(d-1)}^0 (W + i + 1)(1 - p)^{-i+1} \\ &\quad + (W - d + 1) \frac{(1 - p)^{d+1}}{p} \\ &= \sum_{j=0}^{d-1} (W - j + 1)(1 - p)^{j+1} \\ &\quad + (W - d + 1) \frac{(1 - p)^{d+1}}{p} \\ &= \sum_{j=0}^{d-1} (W + 1)(1 - p)^{j+1} - (1 - p)^2 \\ &\quad \times \sum_{j=0}^{d-1} j(1 - p)^{j-1} + (W - d + 1) \frac{(1 - p)^{d+1}}{p} \\ &= (W + 1) \sum_{j=0}^{d-1} (1 - p)^{j+1} \\ &\quad + (1 - p)^2 \frac{\partial}{\partial p} \left( \sum_{j=0}^{d-1} (1 - p)^j \right) \\ &\quad + (W - d + 1) \frac{(1 - p)^{d+1}}{p} \\ &= \frac{(W + 1)(1 - p)(1 - (1 - p)^d)}{p} \\ &\quad + \frac{(1 - p)^2 [dp(1 - p)^{d-1} - 1 + (1 - p)^d]}{p^2} \\ &\quad + (W - d + 1) \frac{(1 - p)^{d+1}}{p} \\ &= \frac{(1 - p)(Wp + 2p - 1 + (1 - p)^{d+1})}{p^2} \end{aligned} \quad (7)$$

Then from Equations (5–7), we get

We can now express the probability that a station transmits in a randomly chosen slot time. As a transmission occurs when the backoff time counter is equal to zero, regardless of the backoff stage, thus,

$$\begin{aligned} \tau &= \sum_{i=-d}^m b_{i,0} = \frac{1 - q^m}{1 - q} b_{1,0} + \frac{1 - p}{p} b_{1,0} \\ &= \frac{1 - q^m}{1 - q} b_{1,0} + \frac{1}{q} b_{1,0} = \frac{1 - q^{m+1}}{(1 - q)q} b_{1,0} \end{aligned} \quad (9)$$

The collision probability is then determined by,

$$p = 1 - (1 - \tau)^{n-1} \quad (10)$$

Note that Equations (8–10) represent a non-linear system in the two unknowns  $\tau$  and  $p$ , which can be solved using numerical techniques. Once  $\tau$  is solved, the throughput can be derived accordingly (neglecting the over-the-air propagation delay):

$$S = \frac{P_s P_{tr} Pk}{(1 - P_{tr})\sigma + P_{tr} P_s T_s + P_{tr} (1 - P_s) T_c} \quad (11)$$

where,

$S$  is the system throughput of MAC layer payload;  $P_s$  is the probability that a transmission occurring on the channel is successful,

$$P_s = \frac{n\tau(1 - \tau)^{n-1}}{1 - (1 - \tau)^n} \quad (12)$$

$P_{tr}$  is the probability that there is at least one transmission in a considered slot time,

$$P_{tr} = 1 - (1 - \tau)^n \quad (13)$$

$Pk$  is the MAC payload size (in bits);  $\sigma$  is the duration of an empty slot time;  $T_s$  is the time the channel is sensed busy because of a successful transmission;  $T_c$  is the time wasted by a collision.

For basic access,  $T_s = T_c = \text{DIFS} + \text{DATA} + \text{SIFS} + \text{ACK}$ ; for RTS/CTS access,  $T_s = \text{DIFS} +$

RTS + SIFS + CTS + SIFS + DATA + SIFS + ACK;  
 $T_c = \text{DIFS} + \text{RTS} + \text{SIFS} + \text{CTS}$ .

The average delay can be estimated as follows. Assume the total throughput given by Equation (11) is fairly shared by the  $n$  contending stations, thus the average packet delay is given by,

$$D = \frac{n \times Pk}{S} = \frac{n \times ((1 - P_{tr})\sigma + P_{tr}P_sT_s + P_{tr}(1 - P_s)T_c)}{P_sP_{tr}} \quad (14)$$

Numerical results are presented in Figure 4 to Figure 6. For 802.11b, the control parameters of MIMLD algorithm,  $CW_{min}$ ,  $CW_{basic}$  and  $CW_{max}$  are, 2, 32 and 1024 respectively. For 802.11a/11g, they are 2, 16 and 1024 respectively.

Figure 4 and Figure 5 present the throughput comparisons of the two algorithms. It can be concluded that MIMLD produces throughput improvements for both 802.11b and 11a/11g, whether the number of active stations is large or small, and whether the packet size is large or small. As for the packet delay, Equation (14) shows that the improvement of the delay is same as that of the throughput.

In particular, when the number of stations is large (say  $> 10$ ) or when it is small (say  $< 3$ ), MIMLD gives significant improvements over the standard algorithm. For instance, in basic access mode, the percentage of improvements for one single station are 24% (802.11b, 1000 bytes packet size), 50% (802.11b, 100 bytes packet size), 24% (802.11a/g, 1000 bytes) and 48% (802.11a/g, 100 bytes). In the case of 60 stations, the improvements are 14% (802.11b, 1000 bytes packet size), 14% (802.11b, 100 bytes packet size),

20% (802.11a/g, 1000 bytes) and 18% (802.11a/g, 100 bytes).

RTS/CTS access is an optional feature in the 802.11 standard and is helpful to overcome the hidden terminal problem. Figure 6 shows that the new backoff algorithm improves the performance of RTS/CTS access as well, although the amount of improvement is not as high as that of basic access.

#### 4.2. Self-Adaptation of the Initial Contention Window

The initial contention window of a packet is the contention window for its first transmission and it should change along with the change of the number of active stations to achieve optimal throughput. However, in the standard backoff algorithm, for every packet, the initial contention window is always set rigidly to  $CW_{min}$ , which does not take into account the current contention intensity. In order to dynamically change the initial contention window, one may come up with a measurement-based scheme. However, as discussed in Section 2, the measurement-based schemes have some drawbacks that render them impractical. Therefore a non-measurement-based scheme is preferable.

Our MIMLD and the standard algorithms are both non-measurement based. They are common in that once the control parameters ( $CW_{min}$ ,  $CW_{basic}$ ,  $CW_{max}$  for MIMLD and  $CW_{min}$ ,  $CW_{max}$  for the standard algorithm) are configured in the system, no further changes are necessary whatever the number of active stations is. However, different from the standard algorithm, in the MIMLD algorithm, the initial contention window of a packet transmission is changed automatically. It is not always  $CW_{min}$  as in the standard algorithm.

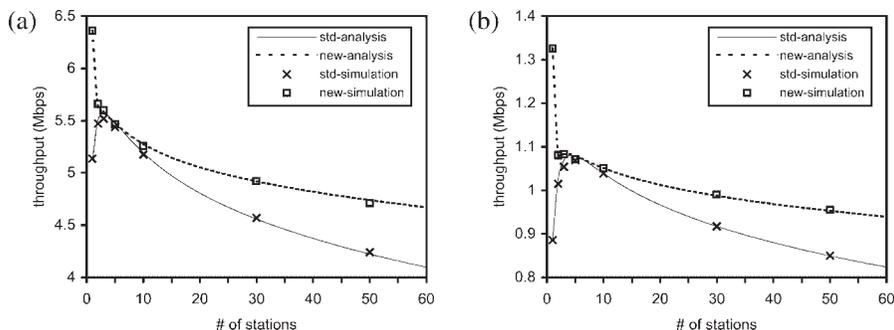


Fig. 4. Throughput comparisons—802.11b. (a) packet size = 1000 bytes. (b) packet size = 100 bytes.

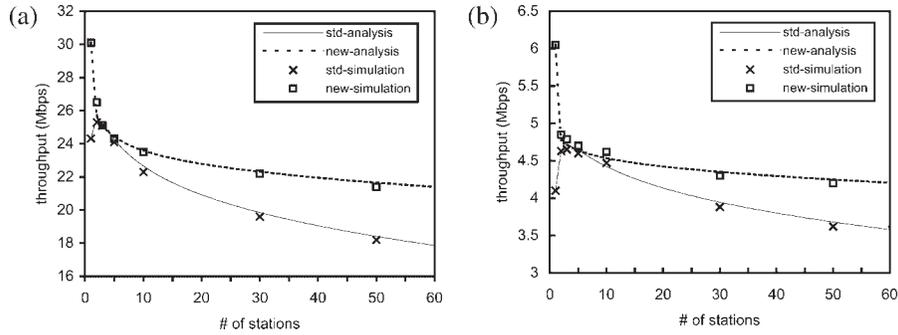


Fig. 5. Throughput comparisons—802.11a/11g. (a) packet size = 1000 bytes. (b) packet size = 100 bytes.

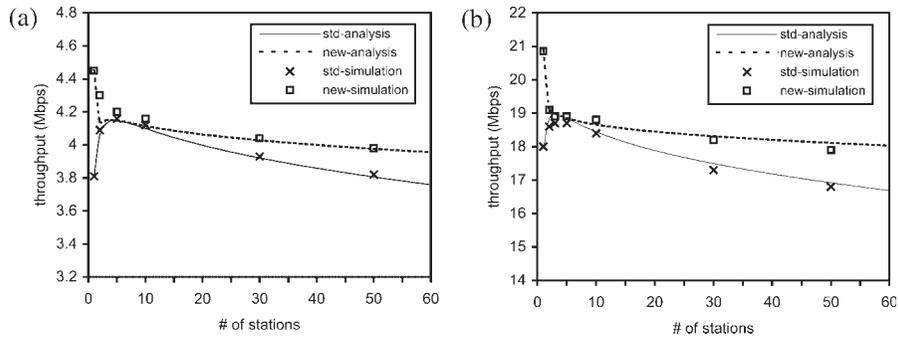


Fig. 6. Throughput of RTS/CTS access (packet size = 1000 bytes). (a) 802.11b. (b) 802.11a/11g.

Table III compares the initial contention windows in the two algorithms when 1000-byte data packets are transmitted over 802.11b by saturated stations. For the standard algorithm,  $CW_{min}$  is 32 and  $CW_{max}$  is 1024. For the MIMLD algorithm, the values for  $CW_{min}$ ,  $CW_{basic}$  and  $CW_{max}$  are 2, 32 and 1024 respectively. All of the parameters are not changed when the number of stations changes. The standard algorithm always uses  $CW_{min}$  (which is 32) as its initial contention window regardless of the number of stations. In contrast, in MIMLD algorithm, the initial contention window adjusts automatically with the change of the number of stations. It can be seen in Table III that the initial contention window in MIMLD is smaller than that in the standard algorithm when the number of stations is small and larger when the number of

Table III. Comparison of initial contention windows.

No. of stations	2	4	6	8	10	20	30	40
Standard algorithm	32	32	32	32	32	32	32	32
MIMLD algorithm	11	25	32	35	38	53	68	120 (average)

stations is large. This explains the reason why the MIMLD algorithm performs better than the standard algorithm.

Figure 7 compares the performance of MIMLD with the standard algorithm with different  $CW_{min}$  values (16, 32 and 64). It is not difficult to conclude

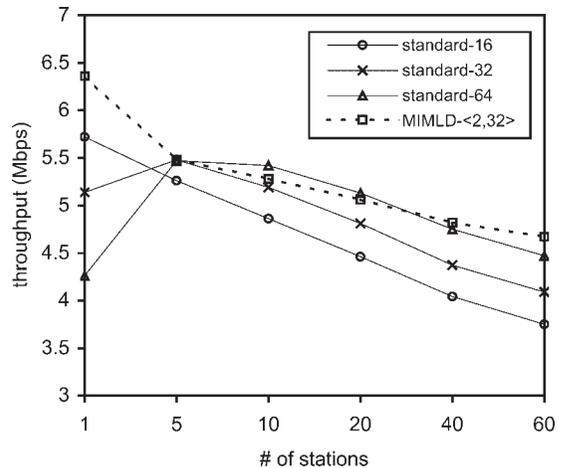


Fig. 7. Comparison of the standard algorithm and the MIMLD algorithm.

that MIMLD performs overall the best among the four cases over a wide range of scenarios.

### 4.3. Fairness and Robustness of MIMLD Algorithm

One may question the fairness of our algorithm by observing that in our algorithm the linear reduction part further reduces the contention window size when the transmission of one station is successful and the window is already below the value of  $CW_{basic}$ . Same as in the standard algorithm, it benefits the successful station. By simulation, we compare the fairness of our algorithm and the standard algorithm quantitatively.

The well-known Jain's fairness index [13] is used. The definition of Jain's index is given by,

$$f(x_1, x_2, x_3, \dots, x_n) = \frac{(\sum_{i=1}^n x_i)^2}{n \sum_{i=1}^n x_i^2} \quad (15)$$

where,  $x_i$  is the throughput of contending flow (station)  $i$ ; and,  $n$  is the number of contending flows (stations).

Table IV gives the measured fairness indices. In the scenarios, saturated stations sending data packets (MAC payload is 1000 bytes) are simulated for 100 s. The results show that there is only a slight fairness impact by the new algorithm.

We then design a simulation scenario in which the number of active stations changes dynamically to examine the robustness of the MIMLD algorithm. The purpose is to demonstrate how our algorithm adapts to the changes in contention intensity. In the scenario being studied, the number of active stations ramps up in steps from 2 to 4, 6, 8, 10, 20, 30 and finally reaches 40. Then the number decreases in steps to 30, 20, 10, 8, 6, 4 and finally back to 2. Each active station operates at the 802.11b physical layer and attempts to send data packets (1000 bytes) in a saturated

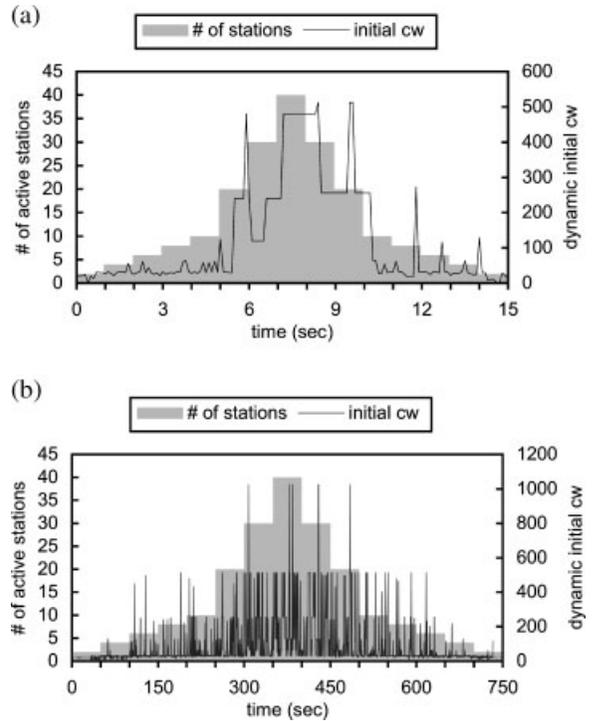


Fig. 8. Dynamics of the sampled initial contention windows in MIMLD. (a) step = 1 s, sampling interval = 0.1 s. (b) step = 50 s, sampling interval = 0.5 s.

manner one after another. Two step sizes, 1 and 50 s, are applied.

Figure 8 shows the sampled initial contention windows. From the figure, when the number of stations is high, the initial contention window tends to be large, and when the number of stations is small, the initial contention window tends to be small. This verifies that our algorithm effectively adapts the contention window according to the contention intensity. Figure 9 shows another scenario in which the number of stations abruptly changes from 5 to 40. The contention windows of two stations are sampled and shown in Figure 9. Within seconds of the change in

Table IV. Comparison of fairness indices.

No. of stations	802.11b				802.11a/g			
	Basic		RTS/CTS		Basic		RTS/CTS	
	Standard	MIMLD	Standard	MIMLD	Standard	MIMLD	Standard	MIMLD
5	0.999	0.999	0.999	0.996	0.999	0.999	0.999	0.999
50	0.994	0.986	0.993	0.984	0.998	0.993	0.998	0.992

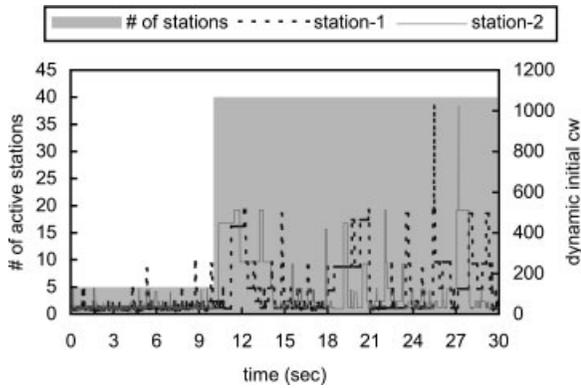


Fig. 9. Reaction of contention window to abrupt change in number of active stations.

the number of active stations, the initial contention window stabilizes to a higher average value.

#### 4.4. Effect of Different Control Parameters in MIMLD Algorithm

We now study the effect of assigning different control parameter values to the MIMLD algorithm. Numerical results in Figure 10 show the effects on system throughput when the control parameters in MIMLD,  $\langle CW_{min}, CW_{basic} \rangle$ , are assigned different values. In the legend, standard-16 and standard-64 stand for the standard algorithm with  $CW_{min} = 16$  and 64 respectively. MIMLD- $\langle 2, 16 \rangle$  stands for MIMLD algorithm with  $CW_{min} = 2$  and  $CW_{basic} = 16$ . MIMLD- $\langle 2, 64 \rangle$  stands for MIMLD algorithm with  $CW_{min} = 2$  and  $CW_{basic} = 64$ . 802.11b physical layer and 1000 bytes data packets are used.

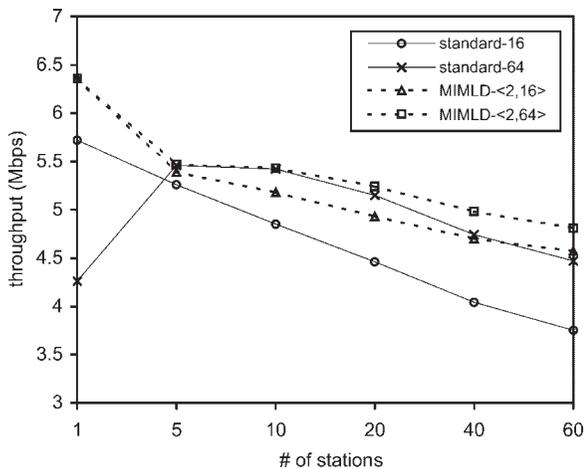


Fig. 10. Effect of MIMLD parameters  $\langle CW_{min}, CW_{basic} \rangle$ .

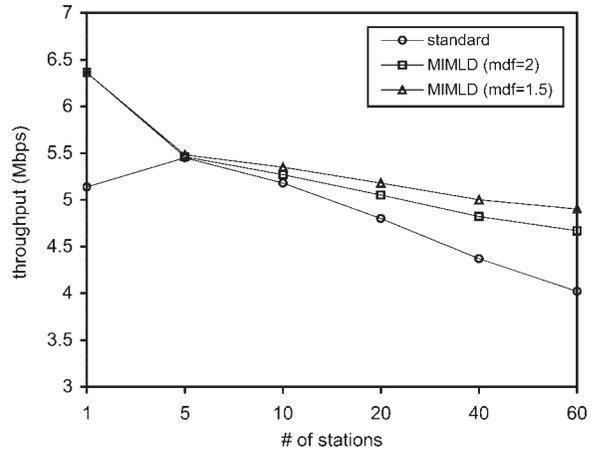


Fig. 11. Effect of different multiplicative decrease factors in MIMLD.

By comparing the results of MIMLD and the standard algorithm, we conclude that the MIMLD algorithm is less sensitive to the contention window settings. The throughput difference between MIMLD- $\langle 2, 16 \rangle$  and MIMLD- $\langle 2, 64 \rangle$  is much smaller than that between standard-16 and standard-64. The insensitivity to the values of  $CW_{min}$  and  $CW_{basic}$  in MIMLD algorithm implies that there is no strong need to adjust them and the values of  $\langle 2, 32 \rangle$  will work well.

In Figure 11, we examine the performance of different multiplicative decrease factors (mdf). Two different values of multiplicative decrease factor are studied, 1.5 and the original 2. In both cases, the contention windows are,  $CW_{basic} = 32$  and  $CW_{min} = 2$ .

Figure 11 shows that by assigning 1.5 to the multiplicative decrease factor, further improvement is obtained when the number of stations is large. But the improvement is less significant than that from the standard algorithm to MIMLD with  $mdf = 2$ .

We would like to leave thorough study of different parameters for future work. In addition, recently the 802.11 standardization committee is actively involved in a new MAC standard of 802.11, 802.11e [14]. In 802.11e, the values of the control parameters, such as  $CW_{min}$ , inter-frame space, transmission opportunity limit, are differentiated for different traffic categories. By applying the same principle, we can extend our new algorithm to its enhanced version—enhanced MIMLD algorithm. As shown in Figure 12, for different traffic categories, different  $CW_{min}$ ,  $CW_{basic}$ ,  $CW_{max}$ , multiplicative increase factor (mif), multiplicative decrease factor (mdf) and linear decrease factor (ldf) can be applied. A study of the performance of enhanced MIMLD algorithm when multiple traffic

- $cw_i \leftarrow \max(cw_i / mdf_i, CWbasic_i)$ , if (succeeds and  $cw_i > CWbasic_i$ ).
- $cw_i \leftarrow \max(cw_i - ldf_i, CWmin_i)$ , if (succeeds and  $cw_i \leq CWbasic_i$ ).
- $cw_i \leftarrow \min\{mif_i \times \max(cw_i, CWbasic_i), CWmax\}$ , if collides.
- $cw_i \leftarrow cw_i$ , if retry limit is reached.

-----  
 $0 \leq CWmin_i \leq CWbasic_i \leq CWmax$   
 $mdf_i$ : multiplicative decrease factor of class  $i$ ;  
 $mif_i$ : multiplicative increase factor of class  $i$ ;  
 $ldf_i$ : linear decrease factor of class  $i$ .

Fig. 12. Service differentiation using the MIMLD algorithm.

categories with different QoS requirements co-exist in one WLAN is left for the future as well.

## 5. Conclusions

In this paper, we have proposed and studied a new self-adaptive contention window adjustment algorithm, MIMLD, for DCF in 802.11 WLANs. Our algorithm has the following advantages over the original and other enhancement algorithms:

- *Effective and flexible*: Compared with previously proposed enhancement algorithms, MIMLD exhibits performance improvements over the original algorithm over a full range of number of active stations. In addition, our algorithm also does not need to assume constant packet size in its optimization procedure.
- *Simple to implement*: Because MIMLD does not need on-line measurements, it can be easily implemented by minor modification of the 802.11 firmware.
- *Robust and responsive*: The simulation results also show that the new algorithm is less sensitive to the initial parameter settings than the standard algorithm. Similarity exists between our MIMLD algorithm and TCP congestion window adjustment algorithm, which has been widely proven robust. Our simulations show that the MIMLD algorithm adjusts to changes in number of active stations within seconds.

In this paper, we mainly target the performance improvement of a single-hop wireless network (either infrastructure mode or ad hoc mode of WLAN), which is currently the dominant application of 802.11 [1,2]. Another potential application of 802.11 products is multi-hop wireless networks. As widely known, there

are severe problems in multi-hop wireless networks [15–17], such as hidden terminal, exposed terminal, self-interference, capture effect, QoS etc. However, the consideration of these problems is outside the scope of this paper.

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## References

1. Gast MS. 802.11 Wireless Networks: The Definitive Guides, O'Reilly, 2002.
2. IEEE Standard for Wireless LAN Medium Access Control (MAC) and Physical Layer (PHY) Specifications, ISO/IEC 8802-11:1999E, August 1999.
3. Bianchi G, Fratta L, Oliveri M. Performance evaluation and enhancement of the CSMA/CA MAC protocol for 802.11 wireless LANs. *IEEE PIMRC* 1996; 392–396.
4. Cali F, Conti M, Gregori E. IEEE 802.11 protocol: design and performance evaluation of an adaptive backoff mechanism. *IEEE Journal on Selected Areas in Communications* 2000; **18**(9): 1774–1786.
5. Peng Y, Wu H, Cheng S, Long K. A new self-adapt DCF algorithm. *IEEE GLOBECOM* 2002; 87–91.
6. Chen Y, Zeng Q, Agrawal DP. Performance analysis and enhancement for IEEE 802.11 MAC protocol. *International Conference on Telecommunications* 2003; 860–867.
7. Wu H, Cheng S, Peng Y, Long K, Ma J. IEEE 802.11 distributed coordination function (DCF): analysis and enhancement. *IEEE ICC* 2002; 605–609.
8. Song N, Kwak B, Song J, Miller LE. Enhancement of IEEE 802.11 distributed coordination function with exponential increase exponential decrease backoff algorithm. *IEEE VTC* 2003 Spring; 2775–2778.
9. Bianchi G. Performance analysis of the IEEE 802.11 distributed coordination function. *IEEE Journal on Selected Areas in Communications* 2000; **18**(3): 535–547.
10. Stevens WR. *TCP/IP Illustrated*, Addison-Wesley, New York, USA, 1994.
11. NS-2, URL <http://www.isi.edu/nsnam/ns>.
12. URL [http://www.tkn.tu-berlin.de/research/802.11e\\_ns2](http://www.tkn.tu-berlin.de/research/802.11e_ns2).
13. Jain R. Throughput fairness index: an explanation. *ATM Forum Contribution* 99-0045, February 1999.
14. IEEE WG, '802.11e D6,' November 2003.
15. Issariyakul T, Hossain E, Kim DI. Medium access control protocols for wireless mobile ad hoc networks: issues and approaches. *Wireless Communications and Mobile Computing* 2003; **3**(8): 935–958.
16. Acharya A, Misra A, Bansal S. High-performance architectures for IP-based multihop 802.11 networks. *IEEE Wireless Communications* 2003; **10**(5): 22–28.
17. Chakrabarti S, Mishra A. Quality of service challenges for wireless mobile ad hoc networks. *Wireless Communications and Mobile Computing* 2004; **4**(2): 129–153.

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